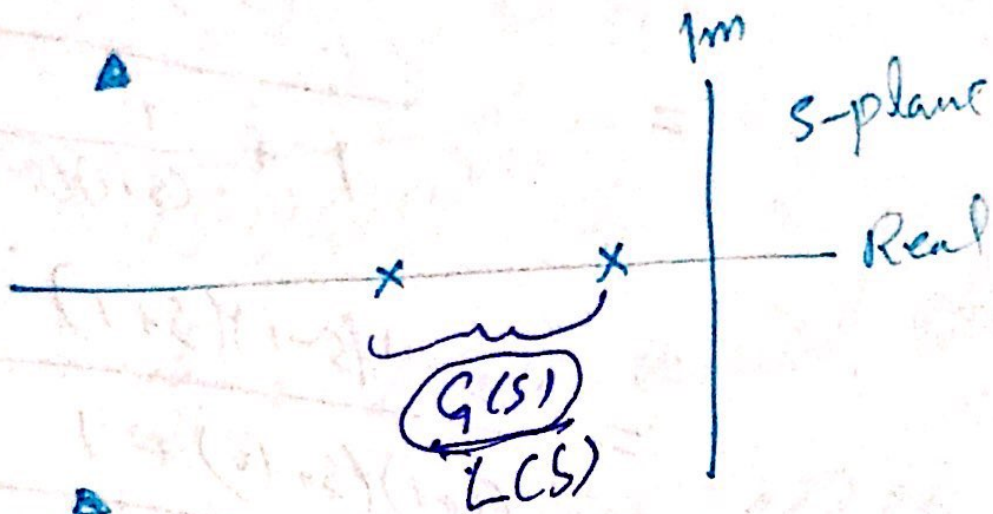
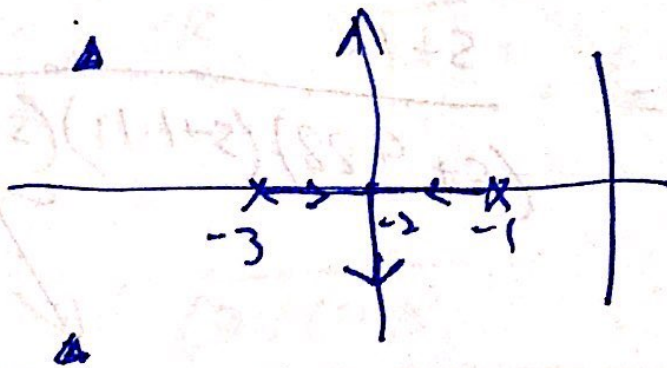


R/L Design Example



desired closed loop pole location.

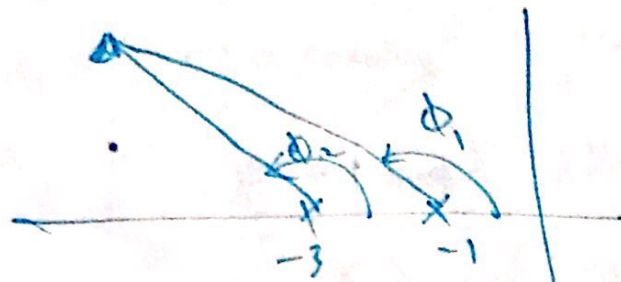
→ To design a controller s.t. \blacktriangle are our c/l pole location.



Can a K_p do this job??

For \blacktriangle to be on the locus we've to satisfy phase and magnitude condition at \blacktriangle .

(4-24)
 Now what is the phase at Δ ??



assume $\phi_1 = 135^\circ$
 $\phi_2 = 120^\circ$

So phase at $\Delta = -135^\circ - 120^\circ = -255^\circ$

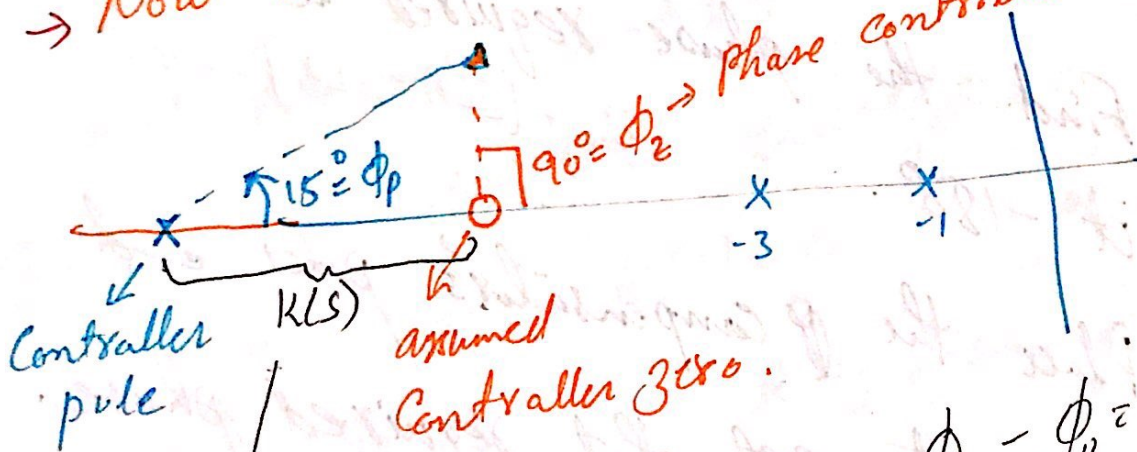
but we need it to -180° .

So we need a phase lead of $-180^\circ + 255^\circ$

$= 75^\circ$, so that $-255^\circ + 75^\circ = -180^\circ$

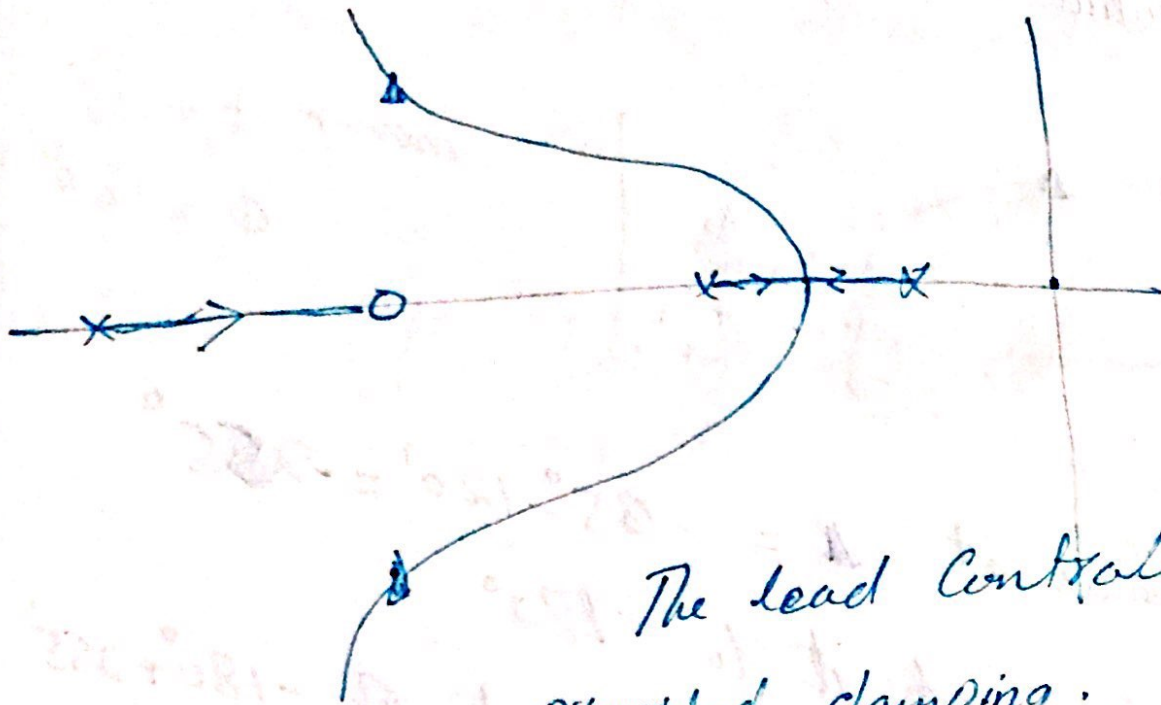
\rightarrow So the controller has to provide 75° phase.

\rightarrow Now we've to build a controller



$\phi_z - \phi_p = 90^\circ - 15^\circ = 75^\circ$

So in this case Δ will be on the locus.



The lead Controller has provided damping.

Steps

① You calculate the phase contribution at the desired location, from the O/L poles and zeros.

$$\sum L\psi_i - \sum L\phi_i$$

② Find the phase required to make it -180° .

$$\theta_{req} + [\sum L\psi_i - \sum L\phi_i] = -180^\circ$$

③ Place the compensator's poles and zero to get that required phase.

④ gain $K_0 = ?$ s.t the locus is stopped at Δ .

Observations \Rightarrow Now we've freedom

to put compensator pole/zero pair
any where we want. ~~It~~ It just has
to ensure the desired phase.

So what will be a good location
for the zero and pole???

① if controller zero is in RHP then
for some high gain a pole will
approach the zero and the sys
will become unstable.

② if the zero is exactly at the
origin then $k(0) = 0$. i.e. in steady
state we are breaking/opening the loop
Although we need $k(0)$ to be bigger for
good steady state tracking.

③ If zero is very near to the origin then ~~it will be distur~~ there will be a very slow pole, disturbing the dominance of the desired pole pair.

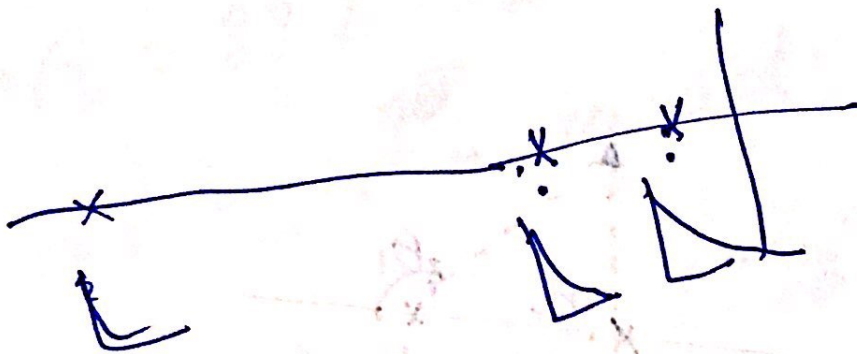
④ If zero is put too much on the left then there will be a very fast pole which may saturate the actuator.

⑤ So a good option is to put the zero in the vicinity ^{of the} ~~of the~~ desired pole.
 \hookrightarrow Real part of the

4-28

Lead Compensator (Design Trade-offs)

- ① Compensator zero not to be in the RHP (Can cause instability)
- ② " " " " on the origin (Large ss error)
- ③ " " " " too close to the origin on the left side (A slow mode in the response is possible)
- ④ " " " " too far on the left (3^{rd} pole may become too fast \Rightarrow actuator saturation)



R/L Design Example

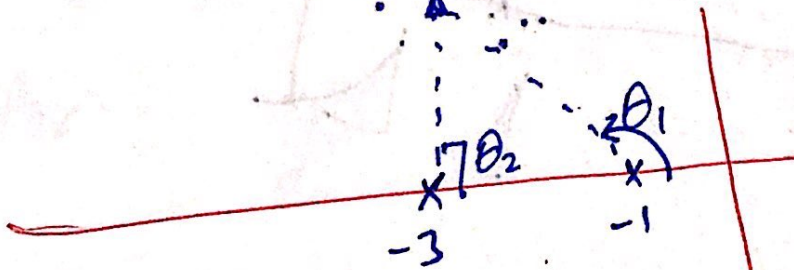
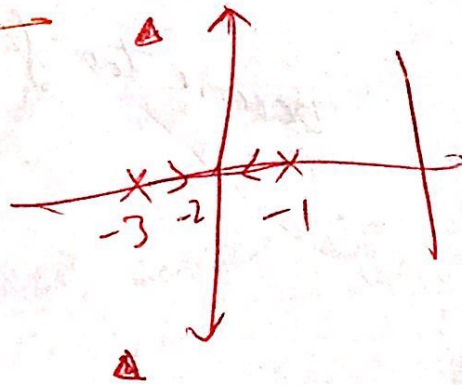
$$G(s) = \frac{1}{(s+1)(s+3)}$$

Desired Specs: (1) c/l poles: $-3 \pm j3$ \rightarrow Transient Resp Specs

(2) $e_{ss} \leq 5\%$ \rightarrow Tracking Specs.

Solution

A simple K_p will not work.



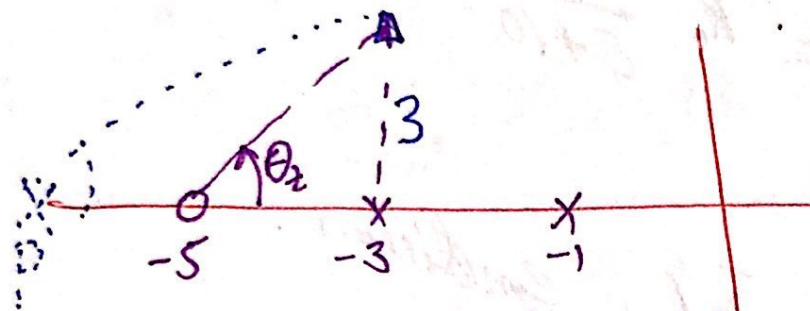
$$\theta_1 = 180^\circ - \tan^{-1}\left(\frac{3}{2}\right) = 123.7^\circ$$

$$\theta_2 = 90^\circ$$

$$\text{Phase} = -90^\circ - 123.7^\circ = -213.7^\circ$$

$\Rightarrow \phi_{\text{lead required}} = -180 + 213.7^\circ = 33.7^\circ$
 to satisfy the phase condition at $\underline{-3 \pm 3j}$

\rightarrow So let $K_{\text{lead}}(s) = \frac{s+5}{s+p}$



$\theta_2 = \tan^{-1} \left(\frac{3}{2} \right) = 56.3^\circ$

$\theta_2 - \theta_p = 33.7^\circ$

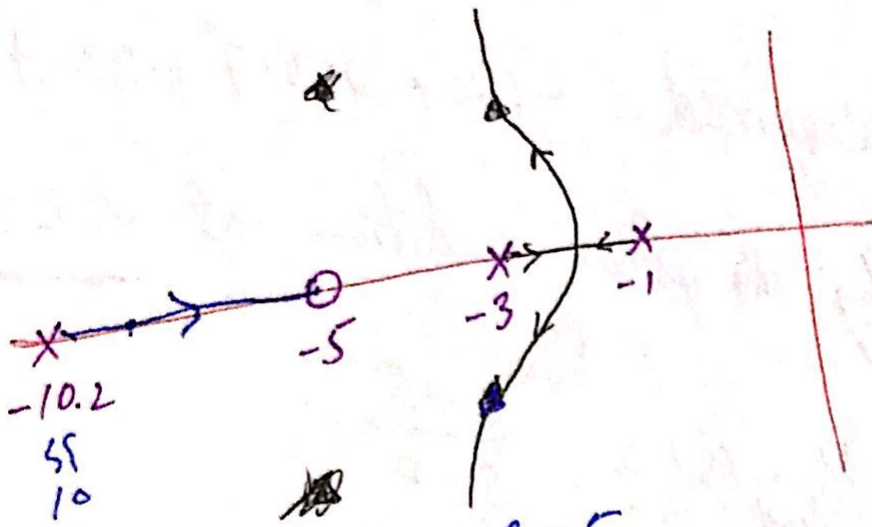
$-\theta_p = 33.7 - 56.3 = -22.6$

$\theta_p = 22.6^\circ$ is required.

$22.6^\circ = \tan^{-1} \left(\frac{3}{y} \right) \Rightarrow \frac{3}{y} = \tan 22.6^\circ = 0.4163$

$y = \frac{3}{0.4163} \approx 7.2$ which means the

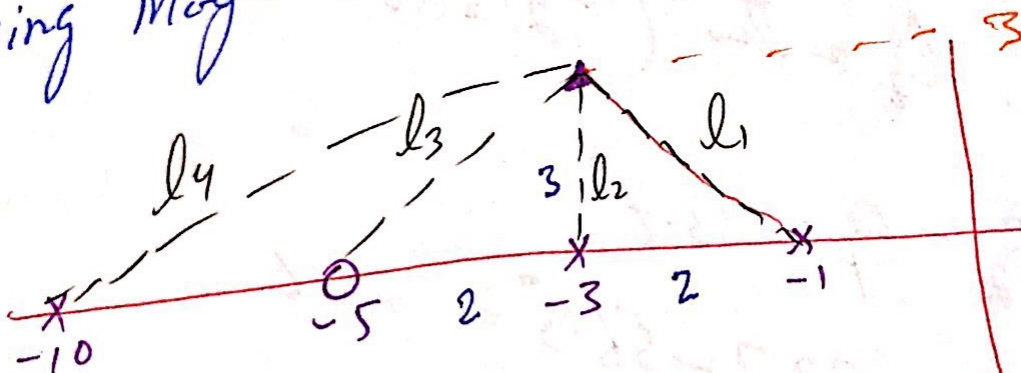
compensator pole is at $-3 - 7.2 = -10.2$



$$So \quad K_{lead} = K_0 \frac{s+5}{s+10}$$

Now $K_0 = ??$

Using Magnitude condition.



$$|K(s)G(s)| = 1$$

$$\left| K_0 \frac{s+5}{s+10} \frac{1}{(s+1)(s+3)} \right| = 1$$

$$\left| K_0 \frac{l_3}{l_4} \cdot \frac{1}{l_1 l_2} \right| = 1$$

$$K \frac{3 \cdot 6}{83.3} = 1$$

$$K_0 = 22.9$$

$$l_1 = \sqrt{2^2 + 3^2} \approx$$

$$l_2 = 3$$

$$l_3 = \approx 3.6$$

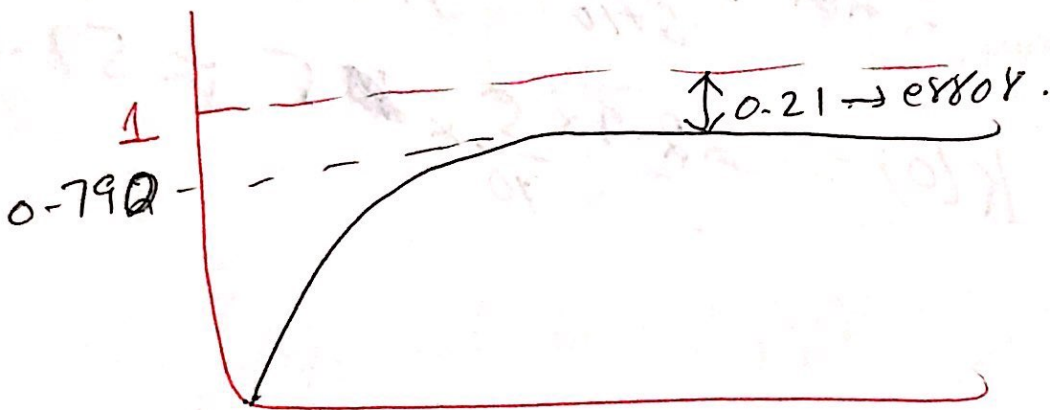
$$l_4 = 7.62$$

$$K_{\text{dead}}(s) = 22.9 \frac{s+5}{s+10}$$

Now the tracking specs.

$$E_{ss} = \frac{1}{1 + K(0)G(0)} = \frac{1}{1 + 22.9 \frac{5}{10} \cdot \frac{1}{3}}$$

$$= \frac{1}{4.82} \approx 0.21$$



We want

$$E_{ss} = \frac{1}{1 + G(0)K(0)} \leq 5\% \leq 0.05$$

$$\approx \frac{1}{1 + 0.33 K(0)} \leq 0.05$$

$$\Rightarrow K(0) \times 0.33 \geq \frac{1}{0.05} - 1 \Rightarrow 20 - 1 = 19$$

$$K(0) \geq \frac{19}{0.33} \geq 57.58$$

So let us choose

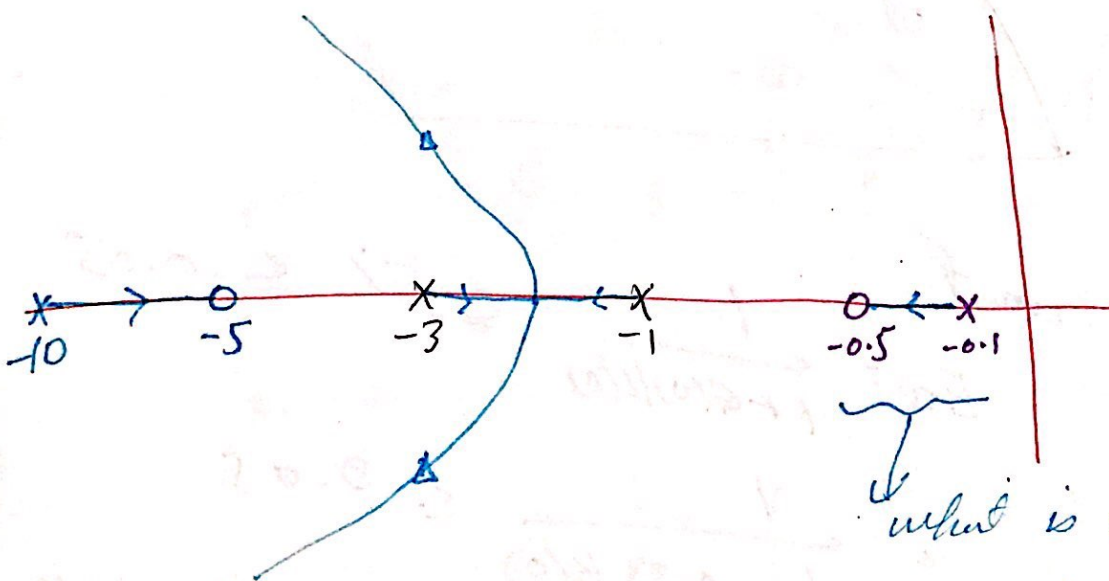
$$K_{lag}(s) = \frac{s+0.5}{s+0.1}$$

↓
to increase the DC gain

$$K(s) = K_{lead}(s) K_{lag}(s)$$

$$= 22.9 \frac{s+5}{s+10} \cdot \frac{s+0.5}{s+0.1}$$

$$K(0) = 22.9 \times \frac{5}{10} \times 5 = 57.25$$



↓
what is it going to do

do we still have the angle conditions satisfied at $-3 \pm 3j$. (K_{lag} behave opposite to K_{lead})

→ If the ~~lead~~ ~~and~~ ~~the~~ lag controller's pole and zero too much far away from the desired location then their respective angles are almost the same. However a lag will still destroy the damping.

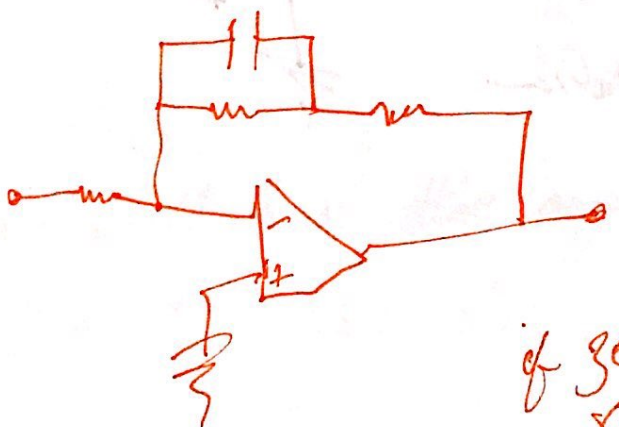
Phase Controller → Provides Phase lead
→ Increases damping
→ Pulls locus leftwards
→ Used for improving transient resp.
→ Move centre of asymptotes to the left.

Practical form of PD

4-35

→ Poles move with different speeds along the locus.

- Gives phase lag
- Reduces damping/Deteriorates transient resp.
- Pushes the locus rightwards
- Used for reducing e_{ss} ,
ref: tracking and dist. rejection.
- Moves the α rightwards



→ Used in place of PI control if zero e_{ss} is not actually required for step size

R/L design guidelines

- ① First try simple K_p
- ② If it doesn't meet the specs, shape the locus by putting in a dynamic compensator (lead or lag)
- ③ For Lead Controller: Lead zero is usually chosen close to the desired CL pole location, lead pole is then put at a factor of 2-25 away depending on the amount of phase lead required.
- ④ Compensator pole & zero are chosen to satisfy the phase condition at the desired CL poles.

⑤ Calculate the K_0 for which the roots are at the desired location.

If more phase lead is required, a double lead can also be used

$$K(s) = K_0 \frac{s+a_1}{s+b_1} \frac{s+a_2}{s+b_2}$$

⑥ For better tracking a lag compensator is used

$$K(s) = K_0 \frac{s+a}{s+b} \quad a > b$$

⑦ The lag zero + pole are usually chosen small enough so as not to disturb the CL pole locations adjusted by the lead controller.

⑧ If too slow the error will decrease very slowly bcz of the lag compensator.

⑨ Final Controller

$$K(s) = K_{lead}(s) K_{lag}(s)$$

$$= K_0 \frac{s + a_{lead}}{s + b_{lead}} \frac{s + a_{lag}}{s + b_{lag}}$$